

TDC1046

Monolithic Video A/D Converter 6-Bit, 25 Msps

Features

- 6-bit resolution
- 1/4 LSB linearity
- · Sample-and-hold circuit not required
- · TTL compatible
- 25 Msps conversion rate
- Selectable output format
- Available in an 18-pin CERDIP
- · Low cost
- · Low analog input capacitance
- · Available per Standard Military Drawing

Applications

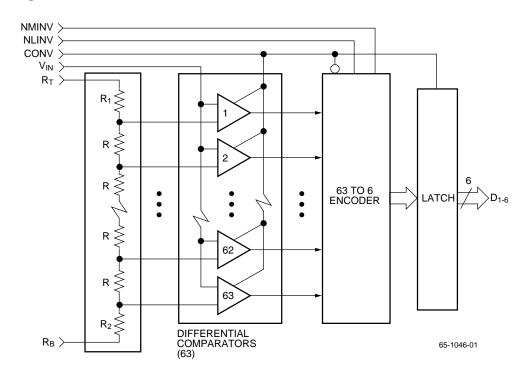
- · Low-cost video digitizing
- · Medical imaging
- · Data acquisition
- · TV special effects
- · Video simulators
- Radar data conversion

Description

The TDC1046 is a 25 Msps (Megasample per second) full-parallel (flash) analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 12.5 MHz into 6-bit digital words. Use of a sample-and-hold circuit is not necessary for operation of the TDC1046. All digital inputs and outputs are TTL compatible.

The TDC1046 consists of 63 clocked latching comparators, encoding logic, and an output buffer register. A single convert signal controls the conversion operation. The unit can be connected to give either true or inverted outputs in binary or offset two's complement coding.

Block Diagram



Functional Description

General Information

The TDC1046 has three functional sections: a comparator array, encoding logic, and output latches. The comparator array compares the input signal with 63 reference voltages to produce an N-of-63 code (sometimes referred to as a "thermometer" code, as all the comparators referred to voltages more positive than the input signal will be off, and those referred to voltages more negative than the input signal will be on). The encoding logic converts the N-of-63 code into binary or offset two's complement coding, and can invert either output code. This coding function is controlled by DC signals on pins NMINV and NLINV. The output latch holds the output constant between updates.

Power

The TDC1046 operates from two supply voltages, +5.0V and -5.2V. The return for ICC, the current drawn from the +5.0V supply, is D_{GND}. The return for I_{EE}, the current drawn from the -5.2V supply, is A_{GND}. All power and ground pins must be connected.

Reference

The TDC1046 converts analog signals in the range $V_{RB} \le V_{IN} \le V_{RT}$ into digital form. V_{RB} (the voltage applied to RB at the bottom of the reference resistor chain) and VRT (the voltage applied to RT at the top of the reference resistor chain) should be between +0.1V and -1.1V. VRT should be more positive than V_{RB} within that range. The voltage applied across the reference resistor chain (V_{RT}–V_{RB}) must be between 0.8V and 1.2V. The nominal voltages are $V_{RT} = 0.00V$ and $V_{RG} = -1.00V$. These voltages may be varied dynamically up to 12.5MHz. Due to variation in the reference currents with clock and input signals, RT and RB should be low-impedance-to-ground points. For circuits in which the reference is not varied, a bypass capacitor to ground is recommended. If the reference inputs are exercised dynamically (as in an Automatic Gain Control circuit), a low-impedance reference source is required.

Controls

Two function control pins, NMINV and NLINV are provided. These controls are for DC (i.e., steady states use. They permit the output coding to be either straight binary or offset two's complement, in either true or inverted sense, according to the Output Coding Table. These pins are active LOW as signified by the prefix "N" in the signal name. They may be tied to VCC for a logic "1" and DGND for a logic "0."

Convert

The TDC1046 requires a CONVert (CONV) signal. A sample is taken (the comparators are latched) within 5ns after a rising edge on the CONV pin. This time is tSTO. Sampling Time Offset. The 63 to 6 encoding is performed on the falling edge of the CONV signal. The coded result is transferred to the output latches on the next rising edge. The outputs hold the previous data a minimum time (tHO) after the rising edge of the CONV signal.

Analog Input

The TDC1046 uses strobed latching comparators which cause the input impedance to vary with the signal level, as comparator input transistors are cut-off or become active. For optimal performance, the source impedance of the driving circuit must be less than 50 Ω . The input signal will not damage the TDC1046 if it remains within the range of VEE to +0.5V If the input signal is at a voltage between VRT and VRB, the output will be a binary number between 0 and 63 inclusive. A signal outside this range will indicate either full-scale positive or full-scale negative, depending on whether the signal is off-scale in the positive or negative direction.

Outputs

The outputs of the TDC1046 are TTL compatible, and capable of driving four low-power Schottky TTL (54/74 LS) unit loads or the equivalent. The outputs hold the previous data a minimum time (tHO) after the rising edge of the CONV signal. Data is guaranteed to be valid after a maximum delay time (tD) after the rising edge of CONV. For optimum performance, 2.2 K Ω pull-up resistors are recommended.

Pin Assignments

Ceramic DIP 18 R_B R_T AGND 17 h D_{GND} DGND 16 13 **NMINV** CONV 15 (MSB) D₁ [5 14 D₆ (LSB) D₂ [6 13 D₅ D_3 12 D₄ 11 NLINV V_{CC} 8 V_{EE} 10 V_{CC} 65-1046-02

Pin Descriptions

Pin Name	Pin Number	Value	Pin Function Description
Power			
Vcc	8, 10	+5.0V	Positive Supply Voltage
VEE	9	-5.2V	Negative Supply Voltage
DGND	3, 16	0.0V	Digital Ground
AGND	17	0.0V	Analog Ground
Reference			
VRT	2	0.0V	Reference Resistor (Top)
VRB	18	-1.0V	Reference Resistor (Bottom)
Controls			
NMINV	4	TTL	Not Most Significant Bit INVert
NLINV	11	TTL	Not Least Significant Bit INVert
Convert			
CONV	15	TTL	Convert
Analog Input			•
VIN	1	0V to -1V	Analog Signal Input
Outputs			•
D ₁	5	TTL	MSB Output
D ₂	6	TTL	
D3	7	TTL	
D4	12	TTL	
D ₅	13	TTL	
D ₆	14	TTL	LSB Output

Output Coding Table¹

	Bina	ry	Two's Co	mplement
Range	True	Inverted	True	Inverted
15.8730mV Step	NMINV=1, NLINV=1	0, 0	0, 1	1, 0
0.0000V	000000	111111	100000	011111
-0.0159V	000001	111110	100001	011110
•	•	•	•	•
•	•	•	•	•
•	•	•	•	•
-0.4921V	011111	100000	111111	000000
-0.5079V	100000	011111	000000	111111
-0.5238V	100001	011110	000001	111110
•	•	•	•	•
•	•	•	•	•
•	•	•	•	•
-0.9841V	111110	000001	011110	100001
-1.0000V	111111	000000	011111	100000

Note:

1. Voltages are code midpoints when calibrated (see Calibration section).

Absolute Maximum Ratings¹

(beyond which the device will be damaged)

Parameter		Min.	Max.	Unit
Supply Volta	ges			
VCC (measur	red to DGND)	-0.5	+7.0	V
VEE (measur	red to AGND)	-7.0	+0.5	V
AGND (meas	ured to DGND)	-0.5	+0.5	V
Input Voltage	es	•	-	
CONV, NMIN	IV, NLINV (measured to DGND)	0.5	+5.5	V
VIN, VRT, VR	B (measured to AGND)	+0.5	VEE	V
V _{RT} (measur	ed to V _{RB})	+1.2	-1.2	V
Output			•	
Applied volta	ge (measured to DGND) ²	-0.5	5.5	V
Applied curre	ent, externally forced ^{3, 4}	-1.0	6.0	mA
Short circuit	duration (single output in high state to ground)		1	sec
Temperature)			
Operating	Case	-55	+125	°C
	Junction		+175	°C
Lead, solderi	ead, soldering (10 seconds) +300			°C
Storage		-65	+150	°C

Notes:

- 1. Absolute maximum ratings are limiting values applied individually while all other parameters are within specified operating conditions. Functional operation under any of these conditions is NOT implied.
- 2. Applied voltage must be current limited to specified range.
- 3. Forcing voltage must be limited to specified range.
- 4. Current is specified as positive when flowing into the device.

Operating Conditions

			Temperature Range					
			Standard			Extended		
Parameter	rs	Min.	Nom.	Max.	Min.	Nom.	Max.	Units
Vcc	Positive Supply Voltage (measured to DGND)	4.75	5.0	5.25	4.5	5.0	5.5	V
VEE	Negative Supply Voltage (measured to AGND)	-4.9	-5.2	-5.5	-4.9	-5.2	-5.5	V
VAGND	Analog Ground Voltage (measured to DGND)	-0.1	0.0	0.1	-0.1	0.0	0.1	V
tPWL	CONV Pulse Width (LOW)	15			15			ns
tpwH	CONV Pulse Width (HIGH)	17			17			ns
VIL	Input Voltage, Logic LOW			0.8			0.8	V
VIH	Input Voltage, Logic HIGH	2.0			2.0			V
loL	Output Current, Logic LOW			4.0			2.0	mA
Іон	Output Current, Logic HIGH			-0.4			-0.4	mA
VRT	Most Positive Reference Input ¹	-0.1	0.0	0.1	-0.1	0.0	0.1	V
VRB	Most Negative Reference Inputs ¹	-0.9	-1.0	-1.1	-0.9	-1.0	-1.1	V
VRT-VRB	Voltage Reference Differential	0.8		1.2	0.8		1.2	V
VIN	Input Voltage	VRB		VRT	VRB		VRT	V

4

Operating Conditions (continued)

		Temperature Range						
		Standard Extended				ed		
Parameter	rs	Min. Nom. Max. Min. Nom. Max.		Max.	Units			
TA	Ambient Temperature, Still Air	0		70				°C
TC	Case Temperature				-55		125	°C

Note:

DC Electrical Characteristics

			Temperature Range				
			Standard Extended		nded		
Param	eter	Test Conditions	st Conditions Min. Max. Min. Max. U		Units		
Icc	Positive Supply Current	V _{CC} = MAX, static ¹		20		25	mA
IEE	Negative Supply Current	VEE = MAX, static ¹			•	'	
		$T_A = 0$ °C to 70 °C		-95			mA
		TA = 70°C		-75			mA
		T _C = -55°C to 125°C				-150	mA
		T _C = 125°C				- 75	mA
IREF	Reference Current	VRT, VRB = NOM		10		15	mA
RREF	Total Reference Resistance	V _{RT} – V _{RB} = MAX	100		66		Ω
RIN	Input Equivalent Resistance	VRT, VRG = NOM, VIN = VRB	40		40		KΩ
CIN	Input Capacitance			30		30	pF
ICB	Input Constant Bias Current	VEE = MAX		105		180	mA
Iμ	Input Current, Logic LOW	VCC = MAX, V _I = 0.5V	•	•	•	•	
		CONV		-0.4		-0.6	mA
		NMINV, NLINV		-0.6		-0.8	mA
lін	Input Current, Logic HIGH	VCC = MAX, VI = 2.4V		50		50	pА
II	Input Current, Max Input Voltage	VCC = MAX, V _I = 5.5V		1.0		1.0	mA
VOL	Output Voltage, Logic LOW	VCC = MIN, IOL = 2 mA		0.5		0.5	V
Voн	Output Voltage, Logic HIGH	VCC = MIN, IOH = MAX	2.4		2.4		V
los	Short Circuit Output Current	VCC = MAX, One pin to ground, one second duration, output HIGH		-30		-30	mA
Cı	Digital Input Capacitance	TA = 25°C, $F = 1MHz$		15		15	pF

Note:

^{1.} VRT must be more positive than VRB, and voltage reference differential must be within specfied range.

^{1.} Worst case, all digital inputs and outputs LOW.

AC Electrical Characteristics

			Temperature Range				
			Stan	dard	Extended		
Parar	meter	Test Conditions	Min	Max	Min	Max	Units
Fs	Maximum Conversion Rate	VCC = MIN, VEE = MIN	25		25		Msps
tsto	Sampling Time Offset	VCC = MIN, VEE = MIN		5		10	ns
tD	Output Delay	VCC = MIN, VEE = MIN, Load 1		30		35	ns
tHO	Output Hold Time	VCC = MAX, VEE = MAX, Load 1	5		5		ns

Timing Diagram

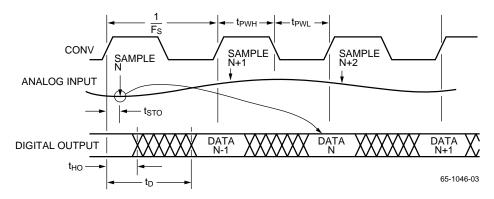
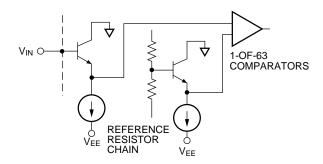


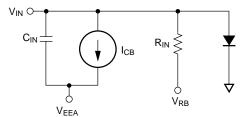
Figure 1. Timing Diagram

System Performance Characteristics

			Temperature Range				
			Stan	dard	Exte	nded	
Paran	neter	Test Conditions	Min	Max	Min	Max	Units
ELI	Linearity Error Integral, Independent	VRT, VRB = Nom		0.4		0.4	%
ELD	Linearity Error Differential			0.4		0.4	%
Cs	Code Size	VRT, VRB = Nom	50	150	50	150	% Nominal
Еот	Offset Error, Top	VIN = VRT		+50		+50	mV
Еов	Offset Error, Bottom	VIN = VRB		-30		-30	mV
Tco	Temperature Coefficient (Offset Voltage)			±20		±20	μV/°C
BW	Bandwidth, Full Power Input		12.5		12.5		MHz
tTR	Transient Response, Full-Scale			10		10	ns
SNR	Signal-to-Noise Ratio	12.5MHz Bandwidth, 2	25Msps	Conve	rsion Ra	ate	
	Peak Signal/RMS Noise	1 MHz Input	42		36		dB
		12.5 MHz Input	40		32		dB
	RMS Signal/RMS Noise	1 MHz Input	33		33		dB
		12.5 MHz Input	31		29		dB
EAP	Aperture Error			30		30	ps

Equivalent Circuits





 C_{IN} IS A NONLINEAR JUNCTION CAPACITANCE V_{RB} IS A VOLTAGE EQUAL TO THE VOLTAGE ON PIN R_{B}

65-1046-04

Figure 2. Simplified Analog Input Equivalent Circuit

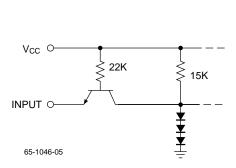


Figure 3. Digital Input Equivalent Circuit

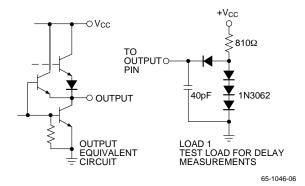


Figure 4. Output Circuits

Applications Discussion

Calibration

To calibrate the TDC1046, adjust VRT and VRB to set the 1st and 63rd thresholds to the desired voltages. In the Block Diagram, note that R₁ is greater than R, ensuring calibration with a positive voltage on RT. Assuming a 0V to -1V desired range, continuously strobe the converter with -0.0079V on the analog input, and adjust VRT for output toggling between codes 00 and 01. Then apply -0.9921V and adjust VRB for toggling between codes 62 and 63. Instead of adjusting VRT, RT can be connected to analog ground and the 0V end of the range calibrated with a buffer offset control. RB is a convenient point for gain adjust that is not in the analog signal path. These techniques are employed in Figure 5.

Typical Interface Circuit

The TDC1046 does not require a special input buffer amplifier to drive the analog input because of its low analog input capacitance. A terminated low-impedance transmission line (< $100~\Omega$) connected to the VIN terminals of the TDC1046 is sufficient if the input voltage levels match those of the A/D converter.

However, many driver circuits lack sufficient offset control, drive current, or gain control. The Typical Interface Circuit

(Figure 5) shows a simple buffer amplifier and voltage reference circuit that may be used with the TDC1046. U2 is a wide-band operational amplifier with a gain factor of -2. A small value resistor, R12, serves to help isolate the input capacitance of the A/D converter from the amplifier output and insure frequency stability. The pulse and frequency response of the buffer amplifier are optimized by variable capacitor C12.

The reference voltage for the TDC1046 is generated by amplifier U3 and PNP transistor Q1 which supplies the reference current. System gain is adjusted by varying R9 which controls the reference voltage level to the A/D converter.

Input voltage range and input impedance for the circuit are determined by resistors R1 and R2. Formulas for calculating values for these input resistors are:

$$R1 \ = \ \frac{1}{\left(\frac{2VR}{Z_{IN}}\right) - \frac{1}{1000}} \quad and \quad R2 \ = \ Z_{IN} - \left(\frac{1000 \ R1}{1000 + R1}\right)$$

where VR is the input voltage range of the circuit, Z_{IN} is the input impedance of the circuit, and the constant 1000 comes from the value of R3. As shown, the circuit is set up for 1Vp-p 75 Ω video input.

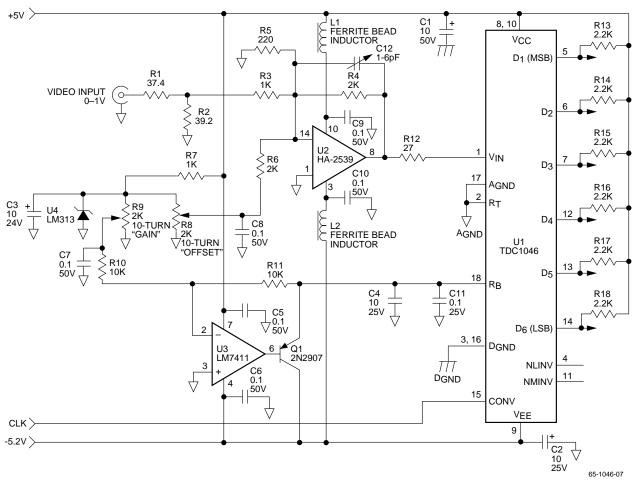


Figure 5. Typical Interface Circuit

Standard Military Drawing

These devices are also available as products manufactured, tested, and screened in compliance with Standard Military Drawings (SMDs). The nearest vendor equivalent product is shown below; however, the applicable SMD is the sole controlling document defining the SMD product.

Standard Military Drawing	Nearest Equivalent Fairchild Product No.	Package
5962-87786-01VA	TDC1046B8V	18 Pin CerDIP

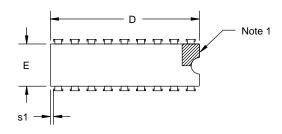
Notes:

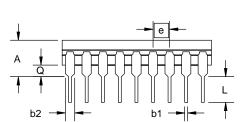
Notes:

Mechanical Dimensions

18-Lead Ceramic DIP

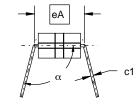
Comple at	Inches		Millimeters		Notes
Symbol	Min.	Max.	Min.	Max.	Notes
Α	_	.200	_	5.08	
b1	.014	.023	.36	.58	8
b2	.045	.065	1.14	1.65	2, 8
c1	.008	.015	.20	.38	8
D	_	.960		24.38	4
Е	.220	.310	5.59	7.87	4
е	.100	.100 BSC		BSC	5, 9
eA	.300	BSC	7.62	BSC	7
L	.125	.200	3.18	5.08	
Q	.015	.070	.38	1.78	3
s1	.005	_	.13	_	6
α	90°	105°	90°	105°	





Notes:

- Index area: a notch or a pin one identification mark shall be located adjacent to pin one. The manufacturer's identification shall not be used as pin one identification mark.
- 2. The minimum limit for dimension "b2" may be .023 (.58mm) for leads number 1, 8, 9 and 18 only.
- 3. Dimension "Q" shall be measured from the seating plane to the base plane.
- 4. This dimension allows for off-center lid, meniscus and glass overrun.
- The basic pin spacing is .100 (2.54mm) between centerlines. Each pin centerline shall be located within ±.010 (.25mm) of its exact longitudinal position relative to pins 1 and 18.
- 6. Applies to all four corner's (leads number 1, 8, 9, and 18).
- 7. "eA" shall be measured at the center of the lead bends or at the centerline of the leads when " α " is 90°.
- 8. All leads Increase maximum limit by .003 (.08mm) measured at the center of the flat, when lead finish is applied.
- 9. Sixteen spaces.



Ordering Information

Product Number	Temperature Range	Screening	Package	PackageMarking
TDC1046B8C	STD-T _A = 0°C to 70°C	Commercial	18 Lead Ceramic DIP	1046B8C
TDC1046B8V	EXT–T _C = -55°C to 125°C	MIL-STD-883	18 Lead Ceramic DIP	1046B8V
5962-87786-01VA	EXT $-T_C = -55^{\circ}C$ to 125°C	Per Standard Military Drawing	18 Lead Ceramic DIP	5962-87786-01VA

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